

PROBLEM-INDEPENDENT ENHANCEMENT OF PML ABC FOR TIME DOMAIN TECHNIQUES IN ELECTRODYNAMICS

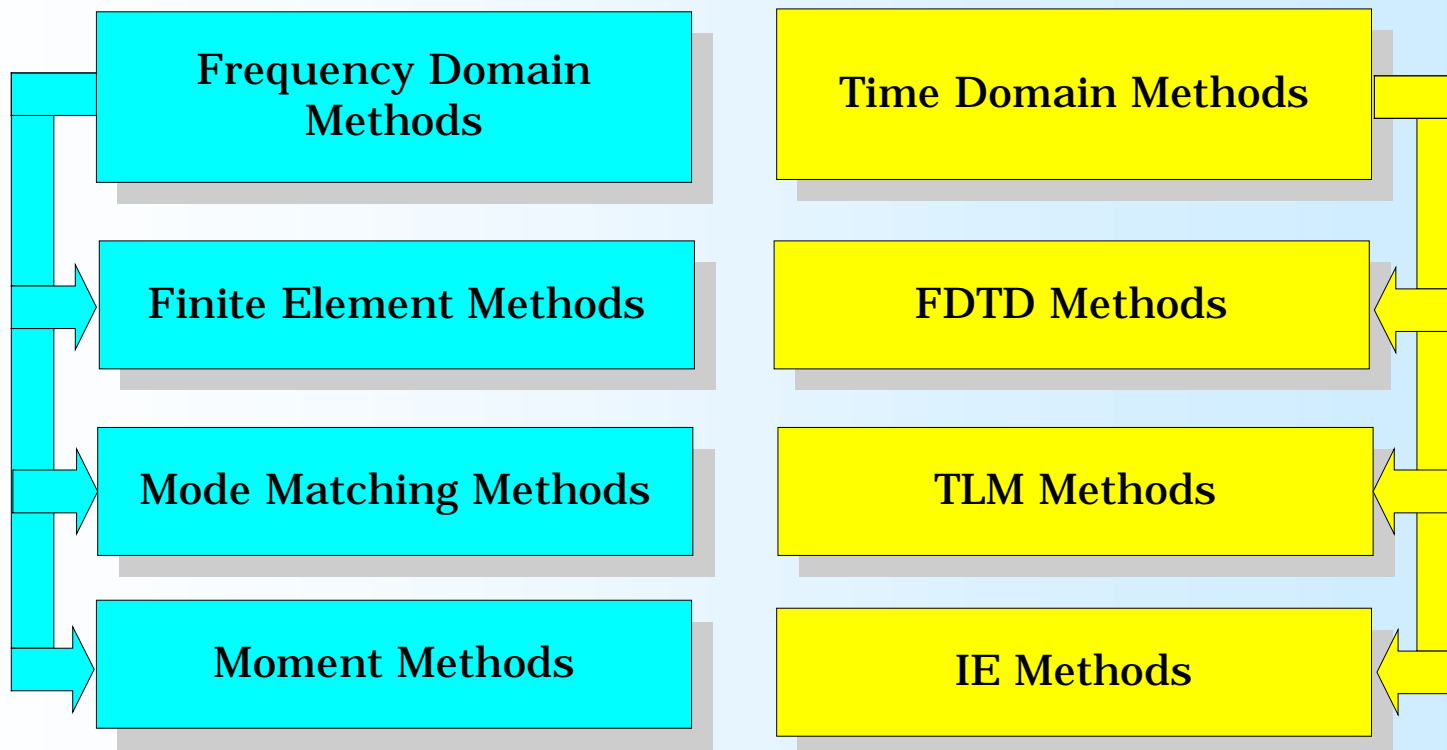
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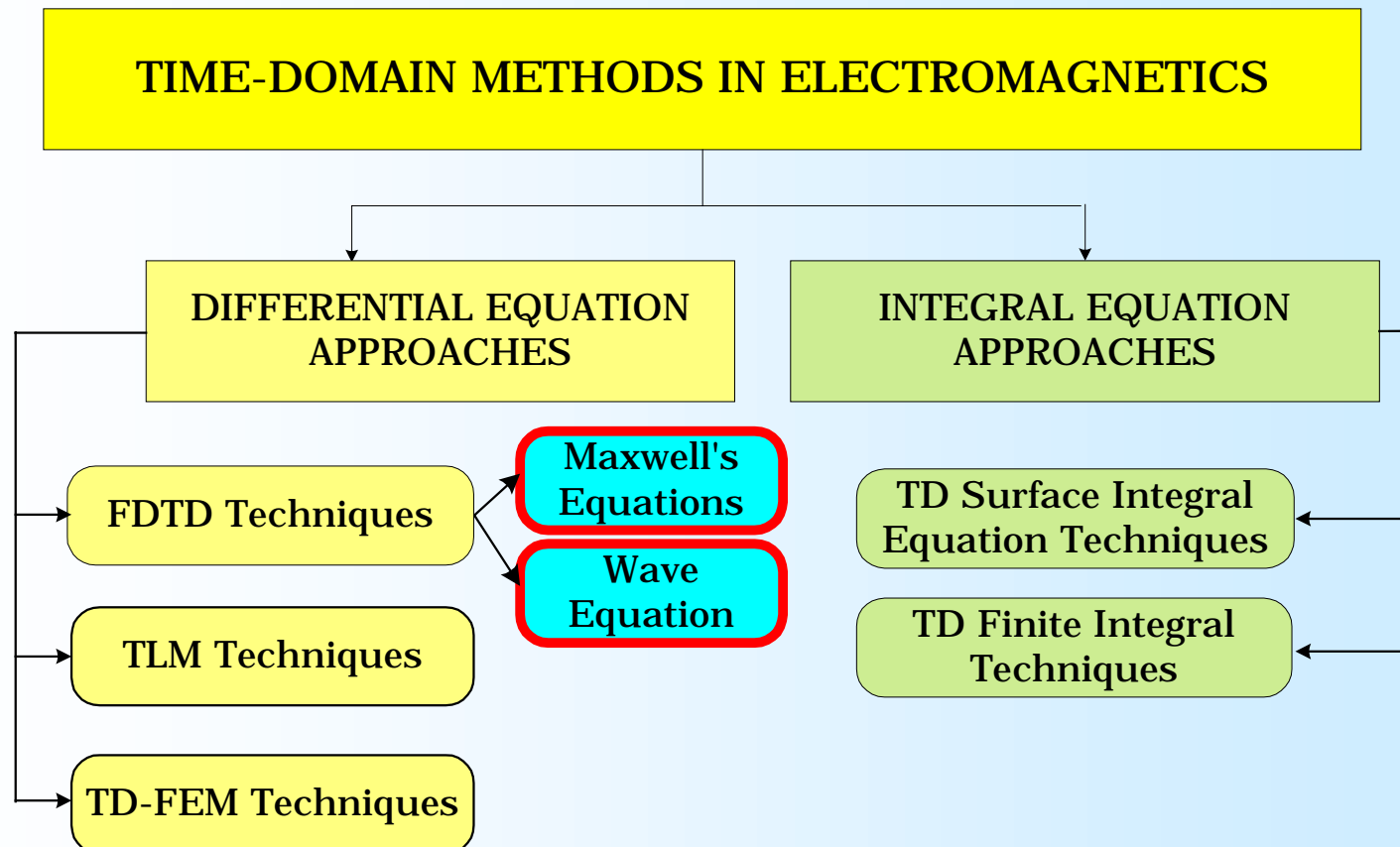
Outline

- Motivation and Objectives
- Background, Challenges and Solutions
- Implementation and Validation
- Conclusion

Classification of Well Known Numerical Techniques for Full Wave EM Field Analysis



Classification of Time-Domain Numerical Techniques for Full Wave EM Analysis



Problem-Independent Enhancement of PML ABC in Electrodynamics

- **New wave-equation techniques in the time domain** emerging:

1. **Vector Potential Technique** N. Georgieva and E. Yamashita (1998), *IEEE Trans. MTT*, vol. 46, No. 4, pp. 404-410.
2. **Wave Potential Technique** N. Georgieva and Y. Rickard (2000), *IEEE MTT-S Int. Symposium Digest*, vol. 2, pp. 1129-1132.

- **Reliable, low-reflection ABCs necessary** for both:

1. Open Problems
2. Guided-Wave Problems

- **Practical applications of the ABCs** in:

1. Microwave Electromagnetics
2. Photonics
3. Any physical problem described by the wave equation or Maxwell's equations in the time domain and requiring reflection-free boundaries

- **Objective:** **Problem-Independent Enhancement of PML ABC for Time-Domain Techniques in Electrodynamics**

Problem-Independent Enhancement of PML ABC in Electrodynamics

• Time-Domain Methods Based on:

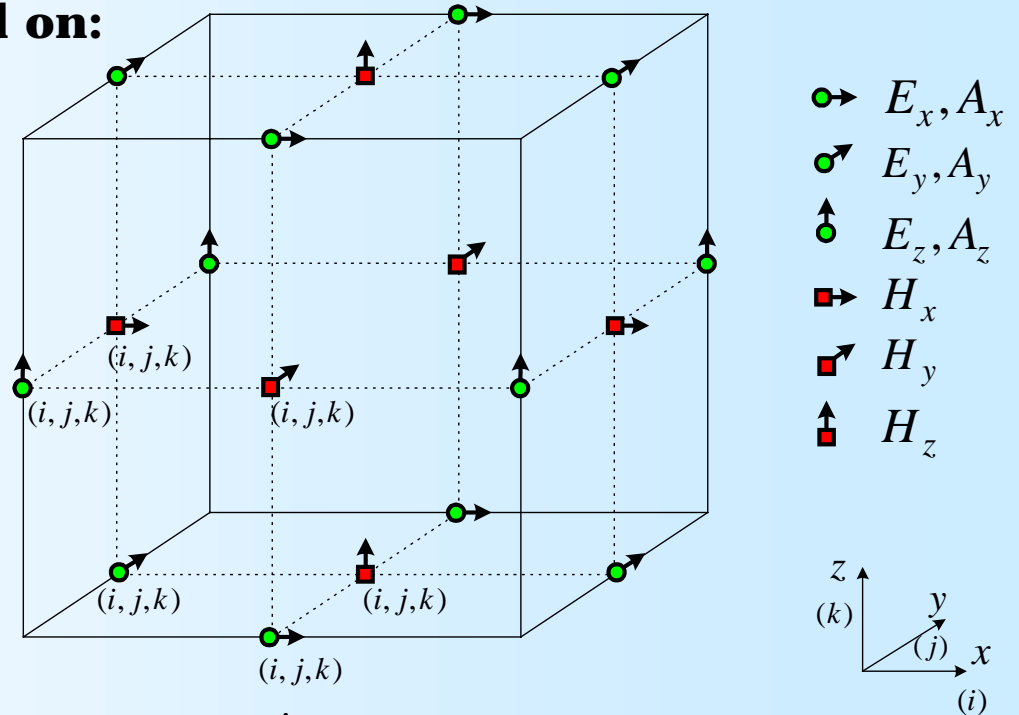
1. Maxwell's curl equations:

$$\nabla \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t} - \sigma_m \vec{H} + \vec{J}_m^i$$

$$\nabla \times \vec{H} = \varepsilon \frac{\partial \vec{E}}{\partial t} + \sigma \vec{E} + \vec{J}^i$$

\Rightarrow **FDTD** (Finite-Difference Time-Domain)

K.S. Yee (1966) , *IEEE Trans. AP*, vol. 14, 1966, pp. 302-307



2. The wave equation in the time domain: \Rightarrow **WETD**

$$\frac{\partial^2 \vec{A}}{\partial t^2} + \left(\frac{\sigma}{\varepsilon} + \frac{\sigma_m}{\mu} \right) \frac{\partial \vec{A}}{\partial t} + \frac{\sigma \sigma_m}{\varepsilon \mu} \vec{A} - \frac{1}{\mu \varepsilon} \nabla^2 \vec{A} - \nabla \left(\frac{1}{\mu \varepsilon} \right) \nabla \cdot \vec{A} - \Phi \nabla \left(\frac{\sigma}{\varepsilon} \right) = \frac{\vec{J}^i}{\varepsilon}$$

N. Georgieva and Y. Rickard (2000), *IEEE MTT-S Int. Symp. Digest*, vol. 2, pp. 1129-1132

Problem-Independent Enhancement of PML ABC in Electrodynamics

- **Berenger's Perfectly Matched Layer (PML) ABC** for the FDTD solution to Maxwell's equations: J.P. Berenger (1994), *J. of Comp. Physics*, vol. 114, pp. 185-200.

1. Split the field components (introducing a new degree of freedom in Yee's FDTD algorithm):

$$\begin{aligned}
 \epsilon_0 \frac{\partial E_x}{\partial t} + \sigma E_x &= \frac{\partial H_z}{\partial y} \\
 \epsilon_0 \frac{\partial E_y}{\partial t} + \sigma E_y &= \frac{\partial H_z}{\partial x} \\
 \mu_0 \frac{\partial H_z}{\partial t} + \sigma^m H_z &= \frac{\partial E_x}{\partial y} - \frac{\partial E_y}{\partial x}
 \end{aligned}
 \Rightarrow
 \left\{
 \begin{aligned}
 \epsilon_0 \frac{\partial E_x}{\partial t} + \sigma_y E_x &= \frac{\partial (H_{zx} + H_{zy})}{\partial y} \\
 \epsilon_0 \frac{\partial E_y}{\partial t} + \sigma_x E_y &= \frac{\partial (H_{zx} + H_{zy})}{\partial x} \\
 \mu_0 \frac{\partial H_{zx}}{\partial t} + \sigma_x^m H_{zx} &= -\frac{\partial E_y}{\partial x} \\
 \mu_0 \frac{\partial H_{zy}}{\partial t} + \sigma_y^m H_{zy} &= -\frac{\partial E_x}{\partial y}
 \end{aligned}
 \right.$$

2. Ensure reflectionless transmission by:

- Tangential wave numbers matching and $\Rightarrow \frac{\sigma_i}{\epsilon_0} = \frac{\sigma_i^m}{\mu_0}; i = x, y, z$
- Normal impedance matching.

3. Gradually increase the direction-dependent PML conductivity.

Problem-Independent Enhancement of PML ABC in Electrodynamics

- From Berenger's PML ABC to Chen's MPML ABC – introduction of PML loss factor (a new degree of freedom):** B. Chen, D.G. Fang and B.H. Zhou (1995), *IEEE MGWL*, vol.5, No. 11, pp. 399-401.

$$\mu_0 \boxed{\kappa_y} \frac{\partial H_{xy}}{\partial t} + \boxed{\sigma_y^m} H_{xy} = - \frac{\partial (E_{zx} + E_{zy})}{\partial y}$$

$$\mu_0 \boxed{\kappa_z} \frac{\partial H_{xz}}{\partial t} + \boxed{\sigma_z^m} H_{xz} = \frac{\partial (E_{yx} + E_{yz})}{\partial z}$$

$$\varepsilon_0 \boxed{\alpha_y} \frac{\partial E_{xy}}{\partial t} + \boxed{\sigma_y} E_{xy} = \frac{\partial (H_{zx} + H_{zy})}{\partial y}$$

$$\varepsilon_0 \boxed{\alpha_z} \frac{\partial E_{xz}}{\partial t} + \boxed{\sigma_z} E_{xz} = - \frac{\partial (H_{yx} + H_{yz})}{\partial z}$$

PML loss factor
for evanescent
wave attenuation

PML conductivity
for propagating
wave attenuation

Problem-Independent Enhancement of PML ABC in Electrodynamics

- **Challenges:**

1. Developing a PML ABC for the wave equation:
 - for time-domain applications in 3 dimensions,
 - for general lossy inhomogeneous media.
2. Developing new PML variable profiles suitable for the wave equation applications.
3. Improving the PML absorber's performance for WETD and FDTD.

- **Solutions:**

1. Use the stretched coordinate approach by mapping the WE into the frequency domain and define auxiliary variables to map it back into the time domain.
2. Introduce a new degree of freedom in the definition of the PML variable profiles – allow growth at different exponent rates.
3. Introduce new types of lossy termination walls for the PML absorbers.

Problem-Independent Enhancement of PML ABC in Electrodynamics

PML for WETD – The Stretched Coordinate Approach

1. W.C. Chew and W.H. Weedon (1994), *MOTL*, vol.7, No. 13, pp.599-604.

$$\nabla_s = \hat{x} \frac{1}{s_x} \frac{\partial}{\partial x} + \hat{y} \frac{1}{s_y} \frac{\partial}{\partial y} + \hat{z} \frac{1}{s_z} \frac{\partial}{\partial z}$$

2. D. Zhou, W.P. Huang, C.L. Xu and D.G. Fang (2001), *IEEE PTL*, vol.13, No.5, pp. 454-456.

Extend D.Zhou's approach to 3 dimensions:

$$\nabla^2 \vec{A} - \mu \epsilon \frac{\partial^2 \vec{A}}{\partial t^2} - (\sigma_m \epsilon + \mu \sigma) \frac{\partial \vec{A}}{\partial t} - \sigma \sigma_m \vec{A} = -\mu \vec{J}$$

$$\nabla_s^2 = \frac{1}{s_x} \frac{\partial}{\partial x} \left(\frac{1}{s_x} \frac{\partial}{\partial x} \right) + \frac{1}{s_y} \frac{\partial}{\partial y} \left(\frac{1}{s_y} \frac{\partial}{\partial y} \right) + \frac{1}{s_z} \frac{\partial}{\partial z} \left(\frac{1}{s_z} \frac{\partial}{\partial z} \right);$$

$$s_\xi = \alpha_\xi + \sigma_\xi / (i\omega \epsilon)$$

$$\xi = x, y, z$$

$$\nabla_s^2 \tilde{\vec{A}} - \mu \epsilon (i\omega)^2 \tilde{\vec{A}} - (\mu \sigma + \epsilon \sigma_m) i\omega \tilde{\vec{A}} - \sigma \sigma_m \tilde{\vec{A}} = -\mu \tilde{\vec{J}}$$

Problem-Independent Enhancement of PML ABC in Electrodynamics

PML for WETD – The Stretched Coordinate Approach, cont'd

Introduce auxiliary variables:

$$\begin{aligned}
 i\omega\tilde{\tilde{X}}_1 &= \frac{1}{s_x} \frac{\partial \tilde{A}}{\partial x}; & i\omega\tilde{\tilde{X}}_2 &= \frac{1}{s_x} \frac{\partial (i\omega\tilde{\tilde{X}}_1)}{\partial x} \\
 i\omega\tilde{\tilde{Y}}_1 &= \frac{1}{s_y} \frac{\partial \tilde{A}}{\partial y}; & i\omega\tilde{\tilde{Y}}_2 &= \frac{1}{s_y} \frac{\partial (i\omega\tilde{\tilde{Y}}_1)}{\partial y} \\
 i\omega\tilde{\tilde{Z}}_1 &= \frac{1}{s_z} \frac{\partial \tilde{A}}{\partial z}; & i\omega\tilde{\tilde{Z}}_2 &= \frac{1}{s_z} \frac{\partial (i\omega\tilde{\tilde{Z}}_1)}{\partial z}
 \end{aligned}
 \Rightarrow
 \begin{aligned}
 \alpha_x \frac{\partial \vec{X}_1}{\partial t} + \frac{\sigma_x}{\epsilon} \vec{X}_1 &= \frac{\partial \vec{A}}{\partial x}; & \alpha_x \frac{\partial \vec{X}_2}{\partial t} + \frac{\sigma_x}{\epsilon} \vec{X}_2 &= \frac{\partial^2 \vec{X}_1}{\partial x \partial t} \\
 \alpha_y \frac{\partial \vec{Y}_1}{\partial t} + \frac{\sigma_y}{\epsilon} \vec{Y}_1 &= \frac{\partial \vec{A}}{\partial y}; & \alpha_y \frac{\partial \vec{Y}_2}{\partial t} + \frac{\sigma_y}{\epsilon} \vec{Y}_2 &= \frac{\partial^2 \vec{Y}_1}{\partial y \partial t} \\
 \alpha_z \frac{\partial \vec{Z}_1}{\partial t} + \frac{\sigma_z}{\epsilon} \vec{Z}_1 &= \frac{\partial \vec{A}}{\partial z}; & \alpha_z \frac{\partial \vec{Z}_2}{\partial t} + \frac{\sigma_z}{\epsilon} \vec{Z}_2 &= \frac{\partial^2 \vec{Z}_1}{\partial z \partial t}
 \end{aligned}$$

$$\mu\epsilon(i\omega)^2 \tilde{\tilde{A}} + (\mu\sigma + \epsilon\sigma_m)i\omega\tilde{\tilde{A}} + \sigma\sigma_m\tilde{\tilde{A}} = i\omega\tilde{\tilde{X}}_2 + i\omega\tilde{\tilde{Y}}_2 + i\omega\tilde{\tilde{Z}}_2 + \mu\vec{J}$$

$$\Rightarrow \mu\epsilon \frac{\partial^2 \vec{A}}{\partial t^2} + (\epsilon\sigma_m + \mu\sigma) \frac{\partial \vec{A}}{\partial t} + \sigma\sigma_m \vec{A} = \frac{\partial \vec{X}_2}{\partial t} + \frac{\partial \vec{Y}_2}{\partial t} + \frac{\partial \vec{Z}_2}{\partial t} + \mu\vec{J}$$

Problem-Independent Enhancement of PML ABC in Electrodynamics

- Problem-Independent Enhancement of the PML ABC:**

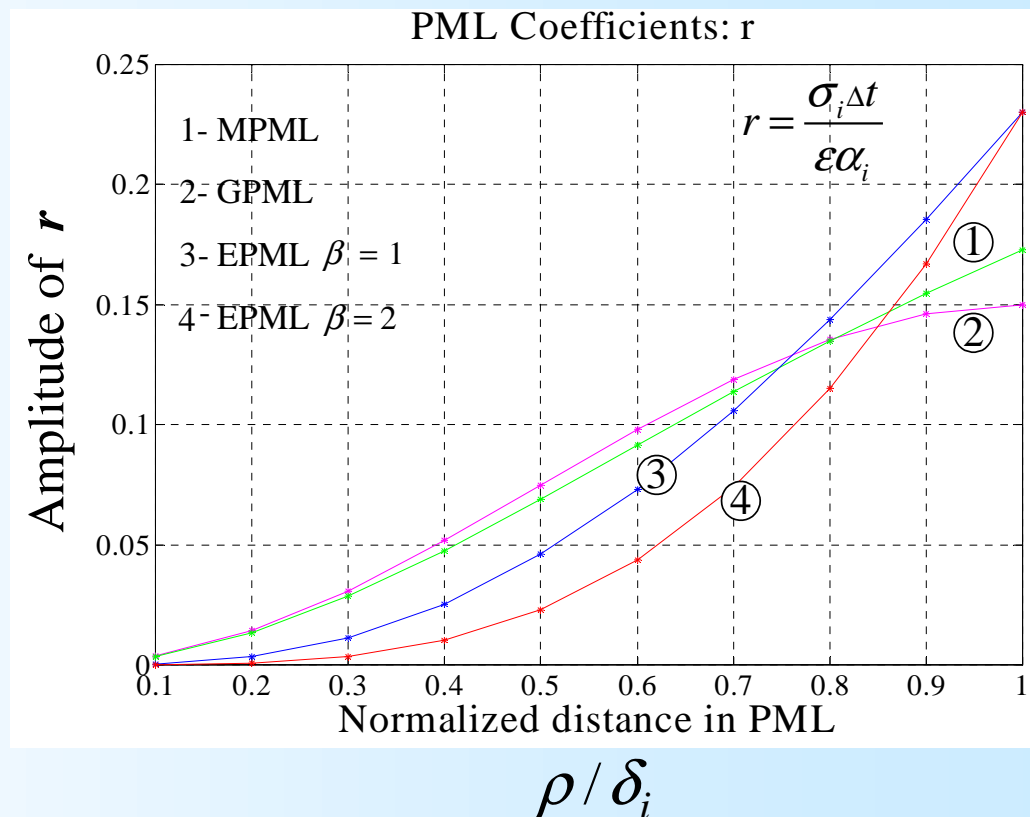
1. Introduction of a new degree of freedom β in the definition of the PML variables:

$$\sigma_i(\rho) = \sigma_{\max} \left(\frac{\rho}{\delta_i} \right)^{n+\beta};$$

$$\beta \in (0, 3]$$

$$\alpha_i(\rho) = 1 + \varepsilon_{\max} \left(\frac{\rho}{\delta_i} \right)^n$$

$$r = \frac{\sigma_i \Delta t}{\varepsilon \alpha_i}$$



Problem-Independent Enhancement of PML ABC in Electrodynamics

- **Problem-Independent Enhancement of the PML ABC, cont'd:**
 2. Introduction of new types of lossy PML termination walls:

The WETD in factored form:

$$(L^2 - \mu\epsilon\nabla^2)\{\vec{A}\} = 0 \quad \Rightarrow \quad \text{1-D version :} \quad L^+ L^- \{\vec{A}\} = 0$$

where

$$L = \frac{\partial}{\partial t} + \frac{1}{\tau}$$

$$\frac{\sigma}{\epsilon} = \frac{\sigma_m}{\mu} = \frac{1}{\tau}$$

$$L^\pm = L \pm v_\xi \partial / \partial \xi$$



$$\frac{1}{v_\xi} \left(\frac{\partial \vec{A}}{\partial t} + \frac{\vec{A}}{\tau} \right) + \frac{\partial \vec{A}}{\partial \xi} = 0$$

(One-way WE
in lossy medium)

Types of Lossy PML Termination Walls:

A. Lossy one-way wave equation

Proposed by: C.M. Rappaport (1996), *IEEE Trans. on Magnetics*, vol. 32, no. 3, pp. 968-974, May 1996.

$$\frac{1}{v_{\xi}} \left(\frac{\partial \vec{A}}{\partial t} + \frac{\vec{A}}{\tau} \right) + \frac{\partial \vec{A}}{\partial \xi} = 0$$

B. Lossy version of Mur's second order ABC

$$\frac{\partial^2 \vec{A}}{\partial t^2} = -\frac{2}{\tau} \frac{\partial \vec{A}}{\partial t} - v \frac{\partial^2 \vec{A}}{\partial \xi \partial t} - \frac{v}{\tau} \frac{\partial \vec{A}}{\partial x} - \frac{1}{\tau^2} \vec{A} + \frac{v^2}{2} \left(\frac{\partial^2 \vec{A}}{\partial \eta^2} + \frac{\partial^2 \vec{A}}{\partial \zeta^2} \right)$$

C. Lossy version of Litva's second order DBC

$$\frac{\partial^2 \vec{A}}{\partial t^2} = -\frac{2}{\tau} \frac{\partial \vec{A}}{\partial t} - (v_{1\xi} + v_{2\xi}) \frac{\partial^2 \vec{A}}{\partial \xi \partial t} - \frac{(v_{1\xi} + v_{2\xi})}{\tau} \frac{\partial \vec{A}}{\partial x} - \frac{1}{\tau^2} \vec{A} - v_{1\xi} v_{2\xi} \frac{\partial^2 \vec{A}}{\partial \xi^2}$$

Implementation and Validation

- **Examples of EPML for FDTD**

1. Dipole in open space
2. Hollow waveguide
3. Microstrip line
4. Optical waveguide with 2-layer AR coating

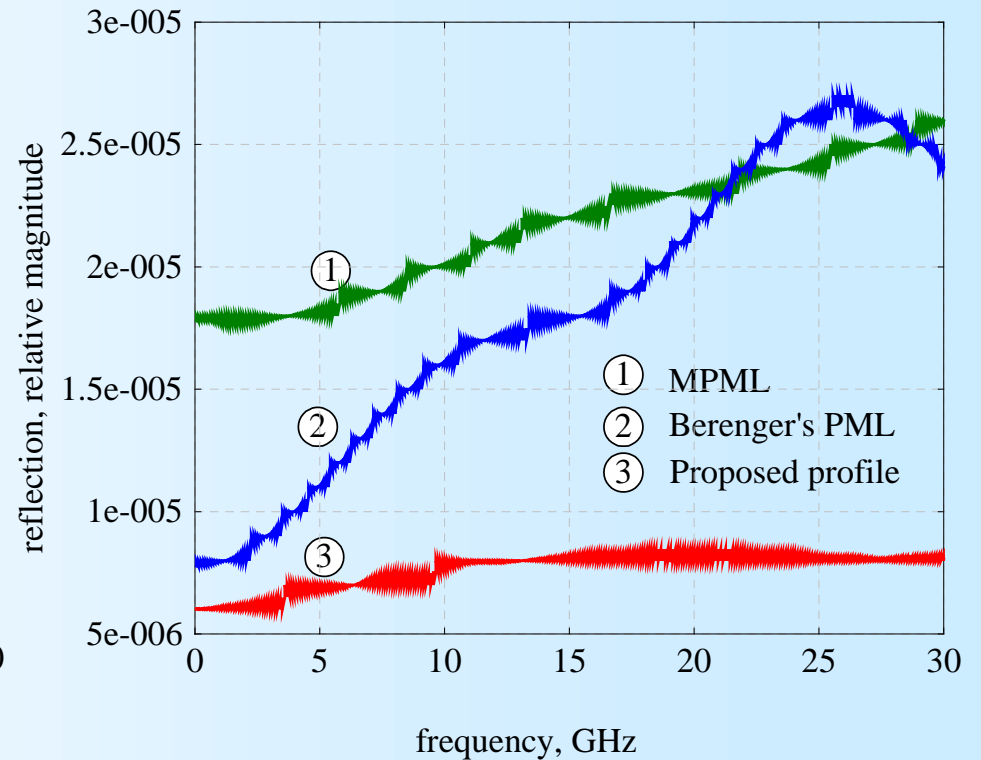
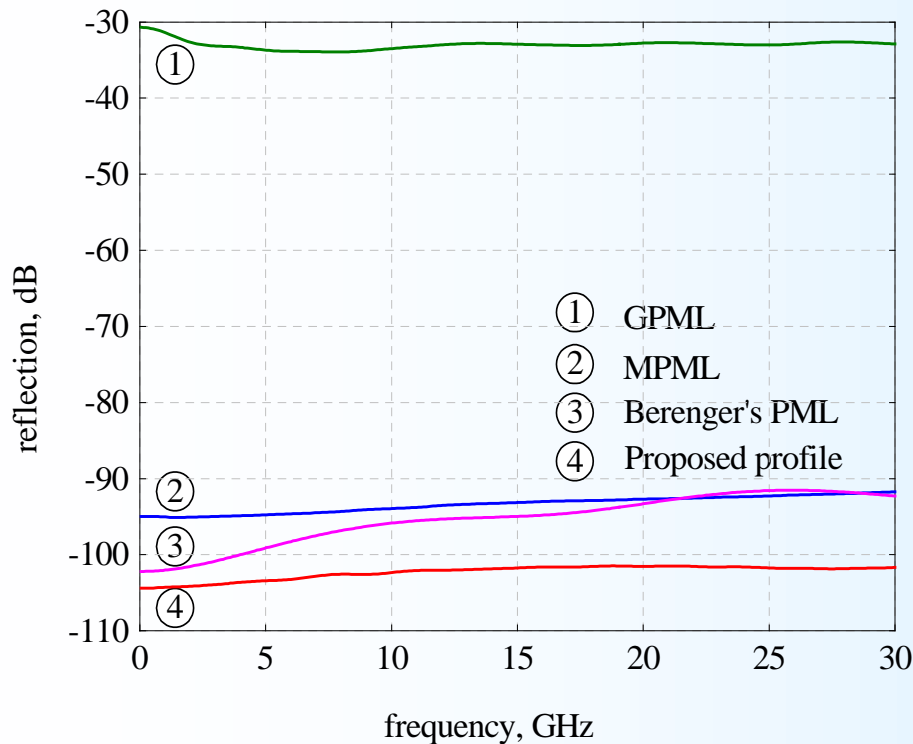
- **Examples of EPML for WETD**

1. Dipole in open space
2. Hollow waveguide
3. Partially filled with dielectric waveguide
4. Optical waveguide with 2-layer AR coating

Reflections:
$$R_{\text{dB}} = 20 \log_{10} \left| \frac{\mathcal{F}\{E_z^{\text{refl}}\}}{\mathcal{F}\{E_z^{\text{inc}}\}} \right|$$

Problem-Independent Enhancement of PML ABC in Electrodynamics

Microstrip line

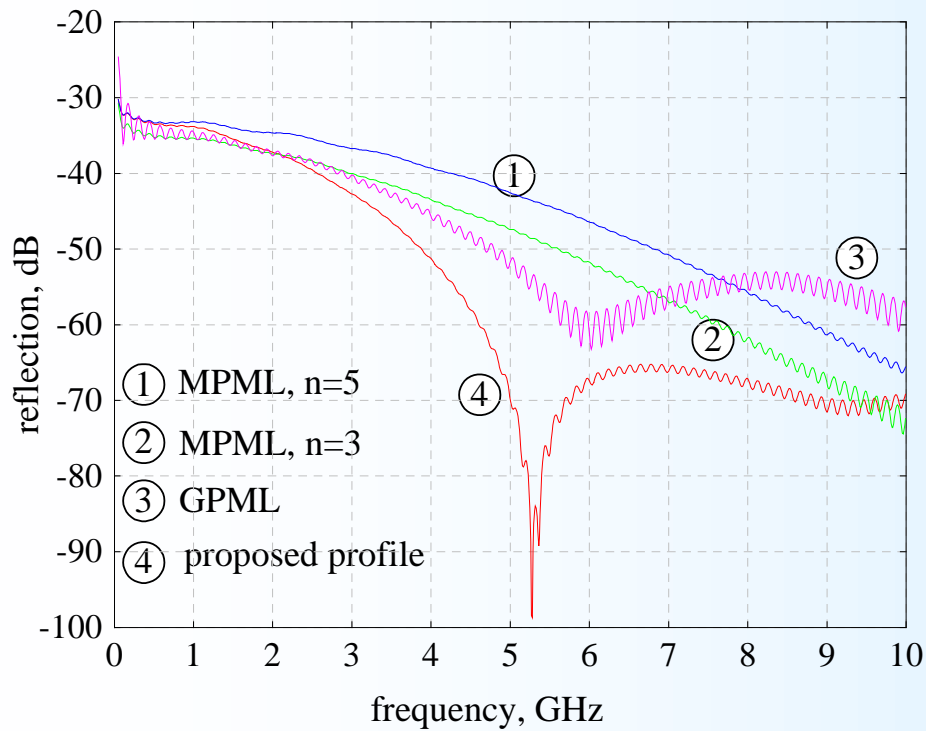


$$N_{\text{PML}} = 8, \quad R_0 = 10^{-3}, \quad \epsilon_{\text{max}} = 1, \quad n = 3$$

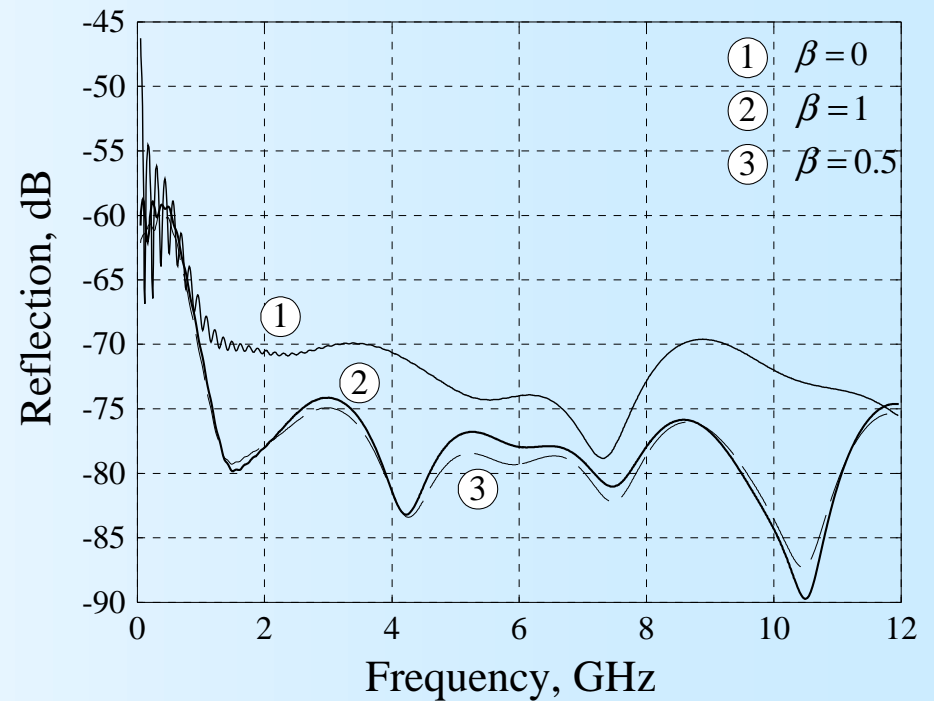
EPML for FDTD

Problem-Independent Enhancement of PML ABC in Electrodynamics

Infinitesimal dipole in open space



$$N_{\text{PML}} = 16, R_0 = 10^{-3}, \epsilon_{\text{max}} = 3, n = 3$$



$$N_{\text{PML}} = 10, R_0 = 10^{-4}, \epsilon_{\text{max}} = 0, n = 2$$

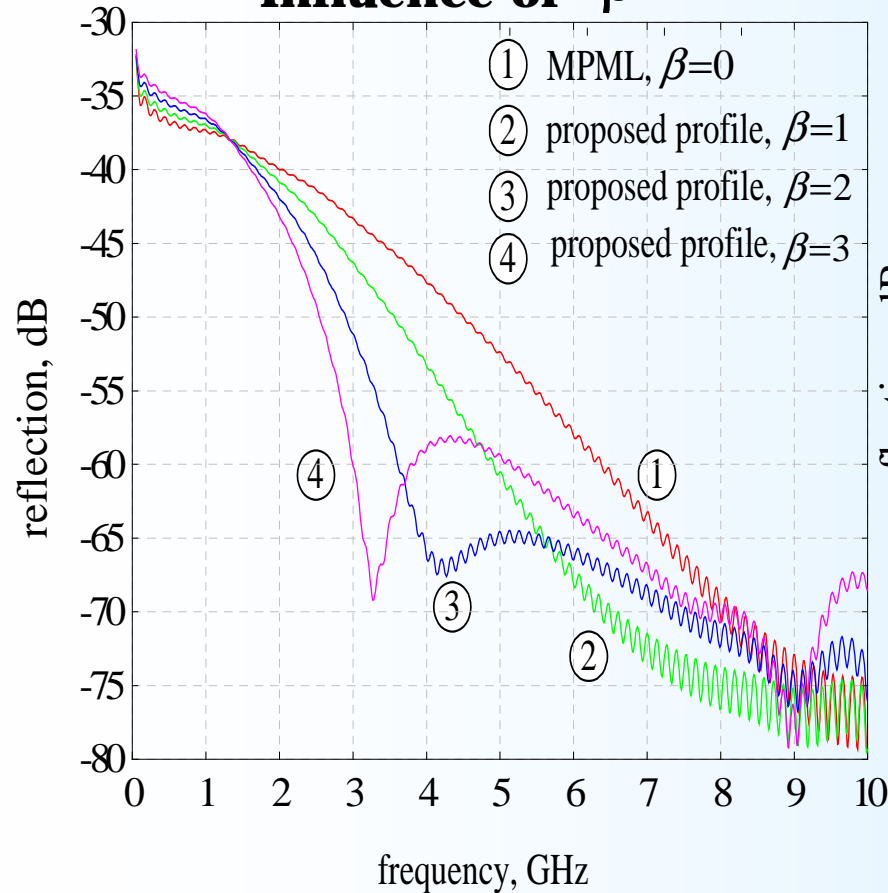
EPML for WETD

EPML for FDTD

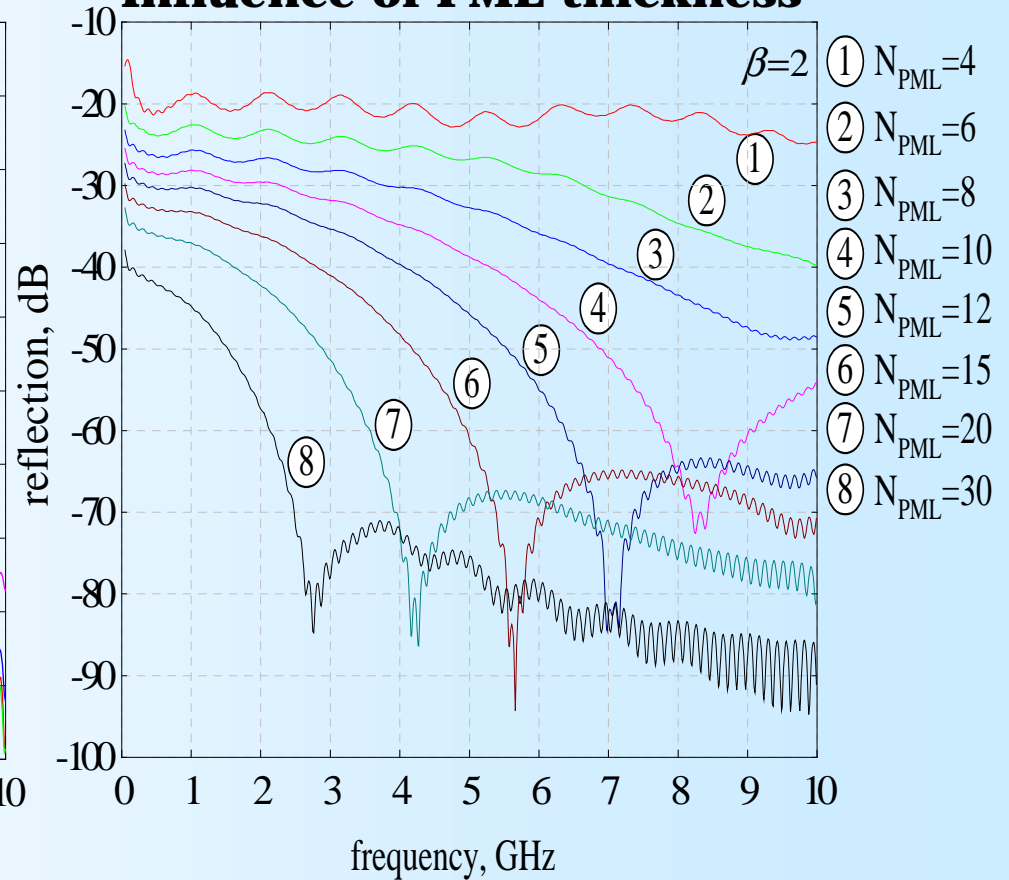
Problem-Independent Enhancement of PML ABC in Electrodynamics

Infinitesimal dipole in open space

Influence of β



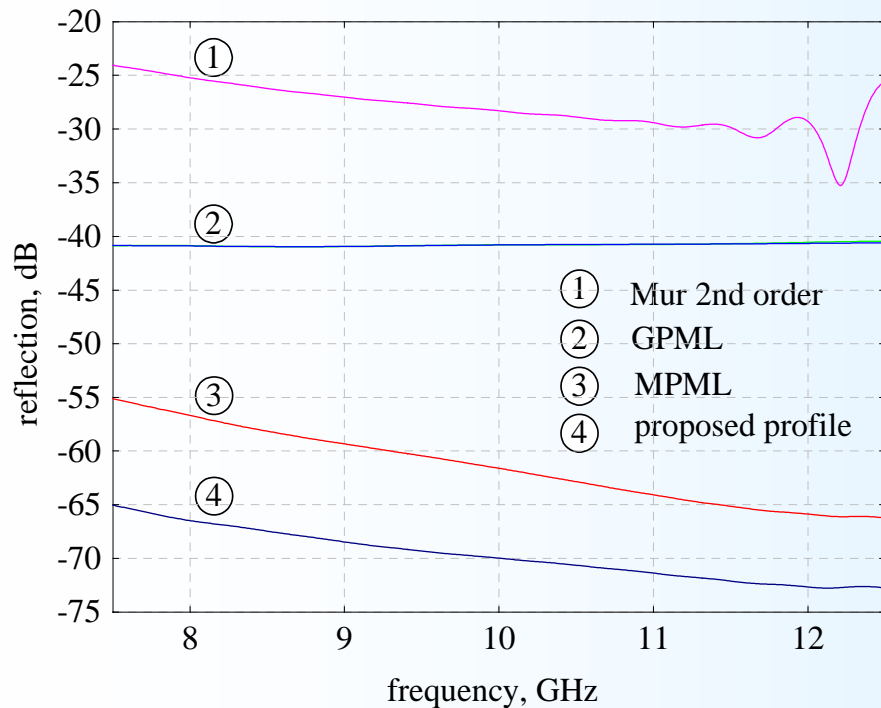
Influence of PML thickness



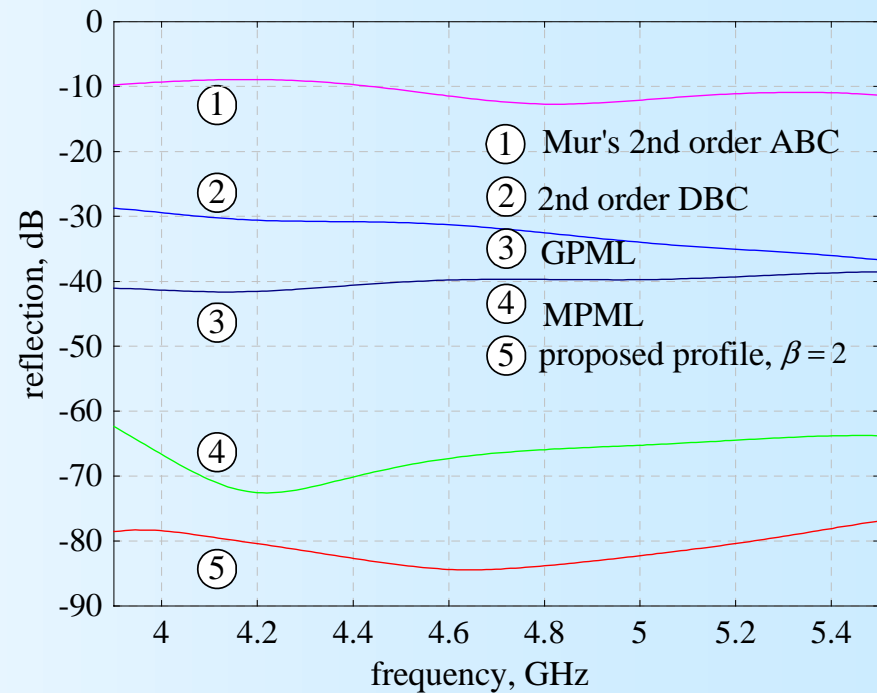
EPML for WETD

Problem-Independent Enhancement of PML ABC in Electrodynamics

Hollow waveguide



Waveguide partially filled with dielectric

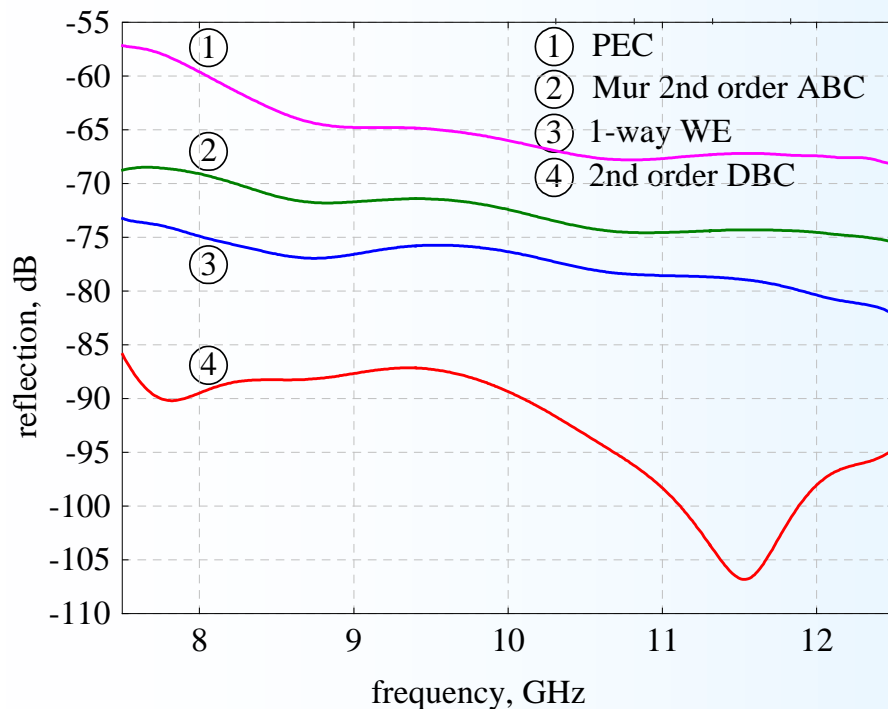


EPML for WETD

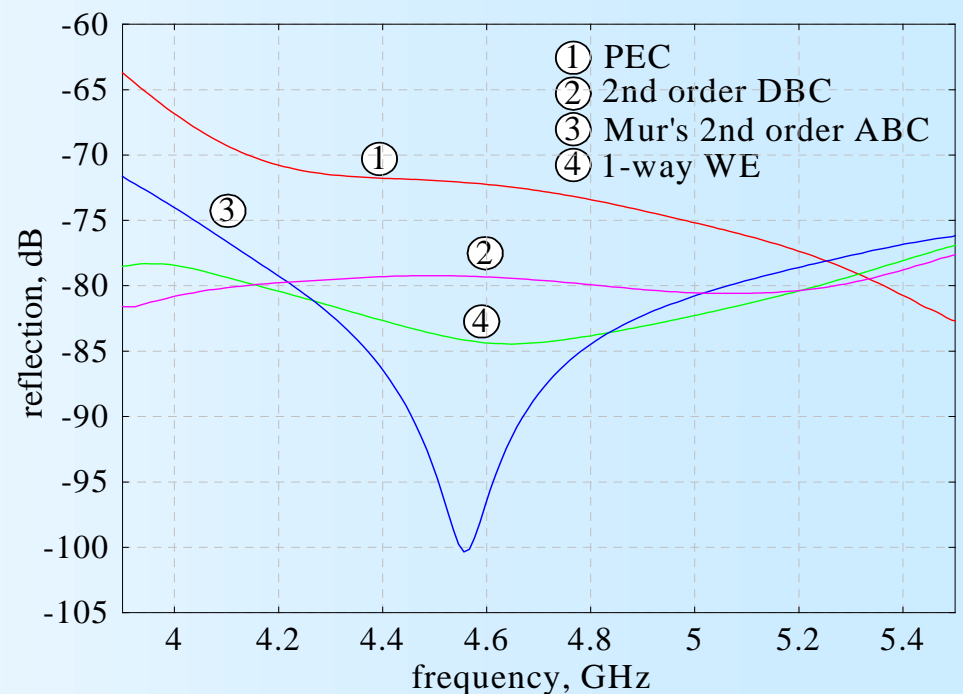
Problem-Independent Enhancement of PML ABC in Electrodynamics

Influence of the Lossy PML Termination Wall

Hollow Waveguide



Waveguide Partially Filled with Dielectric



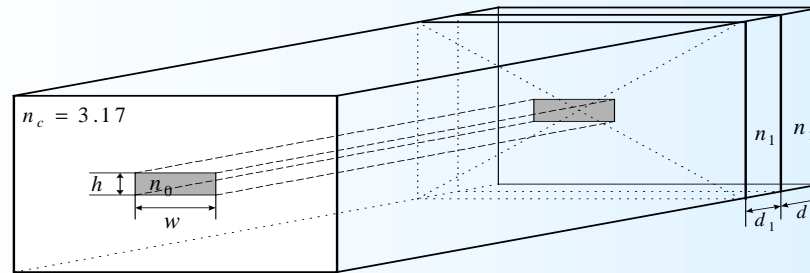
EPML for FDTD

EPML for WETD

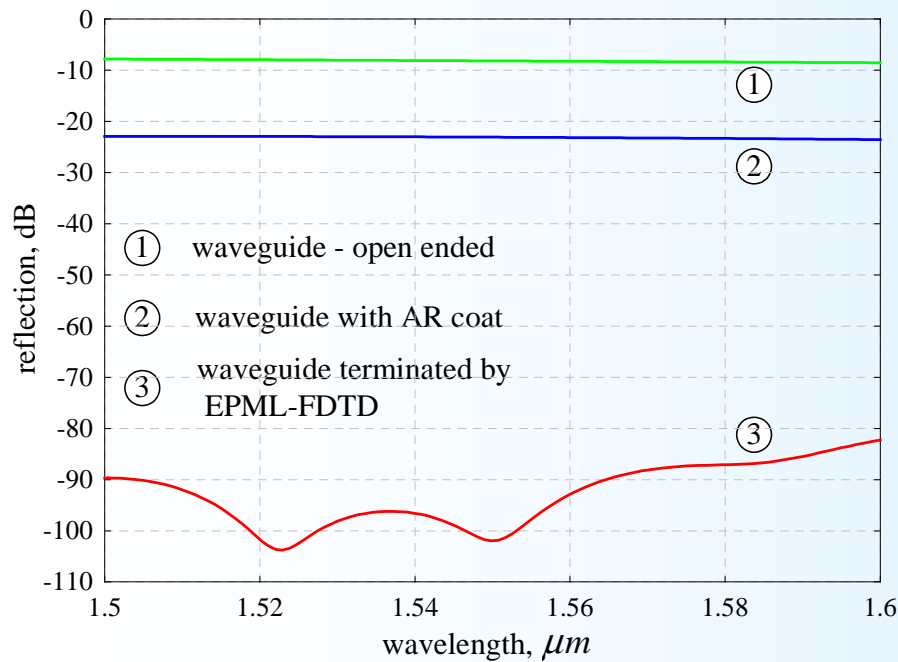
Lossy one-way WE – proposed by C.M. Rappaport (1996), *IEEE Trans. on Magnetics*, vol. 32, no. 3, pp. 968-974, May 1996.

Problem-Independent Enhancement of PML ABC in Electrodynamics

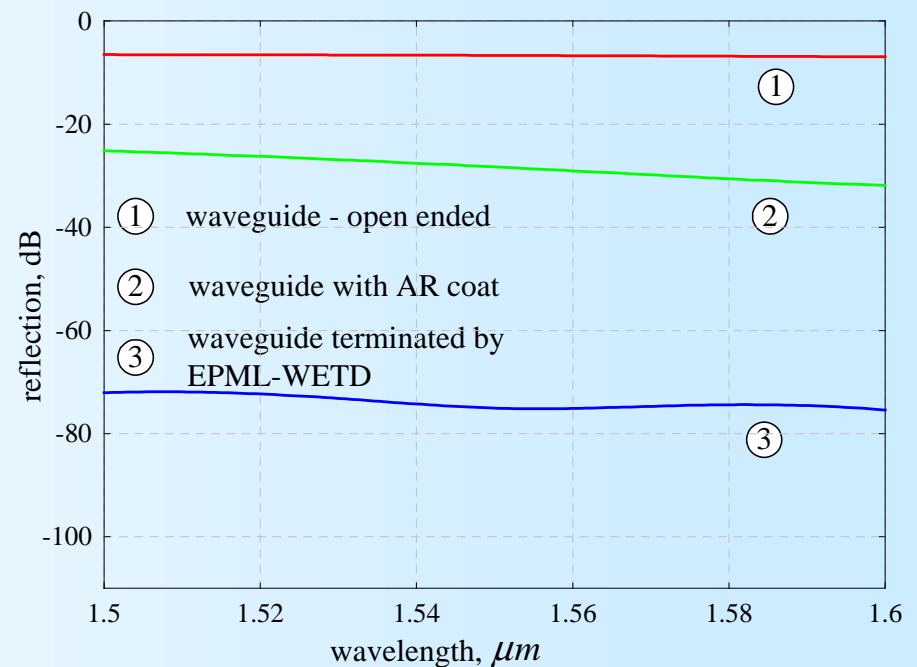
Optical Waveguide with a Two-Layer Antireflection Coating



$$\begin{aligned} d_1 &= 0.166 \mu m & n_1 &= 2.33 \\ d_2 &= 0.267 \mu m & n_2 &= 1.45 \\ w &= 1.6 \mu m & n_0 &= 3.38 \\ h &= 0.1684 \mu m \end{aligned}$$



EPML for FDTD



EPML for WETD

- **Conclusion**

Problem-independent enhancement of PML ABC for the classical Yee's FDTD method and for the WETD method is achieved by the combination of:

1. Introduction of new degree of freedom for PML variable profiles in the spatially polynomial scaling
2. Introduction of modified lossy PML termination walls

Problem-Independent Enhancement of PML ABC in Electrodynamics

- **Suggestions for further development:**
 1. Optimization based on the *combined* influence of all EPML parameters on its performance, including the permittivity of the medium, the user-defined reflection coefficient and the 3rd order discretization error.
 2. Generalization of the EPML ABC for anisotropic and dispersive media.
 3. Application of the EPML ABC to non-linear media.

- **List of Publications Related to the Topic:**

1. Y. Rickard, "Improved Absorbing Boundary Conditions for Time-Domain Methods in Electromagnetics", Ph.D. Thesis, McMaster University, Feb. 2002.
2. Y. Rickard, N. Georgieva and W.-P. Huang, "Application and optimization of PML ABC for the 3-D wave equation in the time domain", in print, *IEEE Trans. on Antennas and Propagation*, Dec 2002.
3. Y. Rickard, N. Georgieva and W.-P. Huang, "A perfectly matched layer for the 3-D wave equation in the time domain", in print, *IEEE Microwave and Wireless Components Letters*, May 2002.
4. Y. Rickard, N. Georgieva and H. Tam, "ABCs for the adjoint problems in the design sensitivity analysis with the FDTD method", submitted to *IEEE Trans. On Microwave Theory and Techniques*.
5. N. Georgieva and Y. Rickard, "The application of the wave potential functions to the analysis of transient electromagnetic fields", *IEEE MTT-S Int. Symposium Digest* (Boston, Massachusetts), June 2000, vol. 2, pp. 1129-1132.
6. N. Georgieva and Y. Rickard, "Time domain modeling of electromagnetic field propagation via wave potentials," *XXVIth General Assembly of the International Union of Radio Science (URSI)*, August 1999, Toronto, Abstracts Digest pp. 178.